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Assistant Commissioner for Patents
Washington, D.C. 20231



PTO-FORM 1082 NEW APPLICATION TRANSMITTAL

Sir:

Transmitted herewith for filing is the X utility patent application of Inventors:

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Entitled: INPUT CIRCUIT AND OUTPUT CIRCUIT

Enclosed are:

1. The Papers Required For Filing Date Under 37 C.F.R. § 1.53(b):

- (23) Page(s) of Specification
- (3) Pages of Claims
- (1) Page of Abstract
- (6) Sheets of Drawings

2. (X) Executed Combined Declaration/Power of Attorney

3. Assignment

- (X) An Assignment (with Recordation Form Cover Sheet) of the invention to:

Matsushita Electric Industrial Co., Ltd.

4. Claim for Priority Under 35 U.S.C. § 120

- () The benefit under 35 U.S.C. § 120 is hereby claimed from the following United States application(s):

5. Claim for Priority Under 35 U.S.C. § 119

- (X) The benefit of priority under 35 U.S.C. § 119 is hereby claimed from the following foreign application(s):

- (1) Japanese Patent Application No. 10-118349
Filed: 28 April 1998

- (X) A Letter submitting a certified copy of each above-identified foreign application(s) is enclosed.

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6. Fee Calculation (claims as filed)

	NUMBER CLAIMS FILED			PRESENT EXTRA	RATE	ADDITIONAL FEE	RATE	ADDITIONAL FEE
TOTAL	8	MINUS	20	= 0	X 9 =	\$	X18 =	\$
INDEP	2	MINUS	3	= 0	X39 =	\$	X78 =	\$
<input type="checkbox"/> FIRST PRESENTATION OF MULTIPLE DEPENDENT CLAIM					+130=	\$	+260=	\$ 260.00
						\$	TOTAL	\$ 260.00

() Amendment cancelling extra claims enclosed

() Amendment deleting multiple dependencies enclosed

7. Small Entity Statement

() A verified statement that this is a filing by a "small entity" under 37 C.F.R. §§ 1.9 and 1.27 is attached.
Filing Fee Calculation (50% of above) \$

8. Fee Payment (enclosed)

(X)	Basic Filing Fee	\$ 760.00
(X)	Claim Fee Calculation Above	\$ 260.00
(X)	Recording Assignment (\$40.00; 37 C.F.R. § 1.21(h)(1))	\$ 40.00
	Total Fees Enclosed	\$1060.00

9. Method of Payment of Fees

(X) Charge Deposit Account No. 13-0203 in
the amount of \$1060.00

10. Other Enclosures: Certified Priority Document(1) Information Disclosure Statement/Cited References(3)

11. Authorization to Credit Overpayment or Charge Additional Fees

(X) The Commissioner is hereby authorized to credit any overpayment, or charge fees under 37 C.F.R. § 1.16 (filing fees) or 37 C.F.R. § 1.17 (application processing fees) which may be required, to Deposit Account No. 13-0203. A duplicate copy of this Transmittal Letter is attached for accounting purposes.

12. Conditional Petition For Extensions of Time

- (X) Petition is hereby made under 37 C.F.R. § 1.136(a) for any extension of time required to ensure that any papers due in the U.S. Patent and Trademark Office during the pendency of this application are timely filed. Please charge any fees due for any extensions of time to Deposit Account No. 13-0203. A duplicate copy of this Transmittal Letter is attached for accounting purposes.

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INPUT CIRCUIT AND OUTPUT CIRCUIT

BACKGROUND OF THE INVENTION

The present invention relates to input circuit and out-
5 put circuit for transferring data within a semiconductor in-
tegrated circuit at a much higher speed.

As the multimedia applications have been expanding in
recent years, improvement of operating performance of a semi-
conductor device has become more and more pressing task to be
10 fulfilled. That is to say, an ideal operating speed of a
semiconductor device has been steeply rising, and desired
power consumption of the device has been drastically falling.
In particular, a system, like an image processor, which is in-
tended to process a large quantity of data at a high speed,
15 strongly needs a semiconductor device operative at an ex-
tremely high speed. Generally speaking, in order to operate a
device at a high speed, data should be transferred within the
device at a high speed, which in turn requires some high-speed
data transfer technique. High-speed-transfer-related technolo-
20 gy includes speeding up the operation of an input/output cir-
cuit and adopting a high-speed transfer standard for a data
bus. Also, input circuits of various types have become popu-
lar lately. Examples of those circuits include: a differen-
tial input circuit for comparing a voltage of a signal
25 received to a reference voltage and amplifying the voltage of

the received signal based on the difference therebetween; and a differential input circuit for receiving differential (or complementary) signals and outputting a single signal.

If an output circuit for outputting a data signal has a push-pull configuration, however, it is very difficult to match its output impedance, or its output current, when each driver transistor outputs H-level data with its output impedance or current when each transistor outputs L-level data. In this specification, "L-level data" and "H-level" data means data with a voltage level defined as logically low (i.e., the voltage level is lower than a certain reference level) and data with a voltage level defined as logically high (i.e., the voltage level is higher than the reference level), respectively. Also, if an output circuit for outputting a data signal is a pull-up output circuit including a pull-up resistor connected to the output thereof, it is hard to equalize a current flowing through the resistor with that flowing through a transistor for outputting the data signal.

In view of the state of the art, the length of a transition interval from L to H level of a data signal is not equal to that of its transition interval from H to L level. Thus, an interval during which the voltage level of such a data signal received by a receiver is defined as logically high judging from the reference voltage is not equal in length to an interval during which that of the data signal is defined

nal is delayed based on at least one of these two transition intervals of the data signal. And the received data signal is latched using at least one of these two types of delayed clock signals and/or the original clock signal.

5 Also, to correct such a difference in length between these two transition intervals of a data signal to be transmitted, the output circuit of the present invention is adapted to control the drivability of a driver thereof outputting the data signal.

10 Specifically, an input circuit according to the present invention includes: delay means for defining a delay time for at least one logical state of a data signal and thereby delaying the clock signal for the delay time defined; and a holding circuit for holding the data signal responsive to the
15 delayed clock signal.

In one embodiment of the present invention, the delay means preferably defines the delay time such that an edge of the clock signal, on which the data signal is intended to be latched and which is included within a transition interval of
20 the data signal, is delayed to a point in time after the transition interval of the data signal is over.

In this particular embodiment, the delay means may include: a comparator for comparing the edge of the clock signal, on which the data signal is intended to be latched, to
25 at least one of leading and trailing edges of the data sig-

nal; and a delay circuit for defining the delay time based on a result of comparison performed by the comparator.

In an alternate embodiment, the delay means may include: a comparator for comparing the edge of the clock signal, on which the data signal is intended to be latched, to leading and trailing edges of the data signal; a first delay circuit for defining the delay time for a logically high state of the data signal based on a result of comparison, performed by the comparator, between one of the leading edges of the data signal and the edge of the clock signal; a second delay circuit for defining the delay time for a logically low state of the data signal based on a result of comparison, performed by the comparator, between one of the trailing edges of the data signal and the edge of the clock signal; and a selector for selecting the delay time defined by the first delay circuit when the data signal is in the logically high state or the delay time defined by the second delay circuit when the data signal is in the logically low state.

In another embodiment, the delay circuit may define the delay time based on the result of comparison performed by the comparator and a setup time for correctly latching the data signal.

An output circuit according to the present invention includes: a driver including a plurality of devices outputting a data signal, the total drivability of the devices being

controllable; and a controller, responsive to a signal representing a transition interval length of the data signal, for increasing or decreasing the drivability of the driver.

In one embodiment of the present invention, the controller preferably receives the signal, representing the transition interval length of the data signal, from an input circuit to which the data signal is output from the output circuit.

The input circuit of the present invention can prevent erroneous latching. For example, suppose the transition interval of a data signal from H into L level is relatively long. In such a case, even though an edge of the original clock signal, on which the data signal is usually latched, is included within the transition interval, the original clock signal is delayed for a relatively long amount of time defined for the logically state of the data signal. Accordingly, an edge of the delayed clock signal, on which the data signal is actually latched, is located at a point in time after the transition of the data signal into the L level is over. As a result, the L-level data signal can be latched correctly with a lot more certainty.

The output circuit of the present invention can also prevent erroneous latching. For example, suppose the transition interval of a data signal from H into L level is relatively long. In such a case, even though an edge of the

clock signal, on which the data signal is usually latched, is included within the transition interval, the transition interval can be shortened, because the drivability of the driver is increased. Thus, the edge of the clock signal, on which the data signal is actually latched, is located at a point in time after the shortened transition interval of the data signal into the L level is over. As a result, the L-level data signal can be latched correctly with a lot more certainty.

BRIEF DESCRIPTION OF THE DRAWINGS

Figure 1 is a block diagram illustrating an internal configuration of an input circuit according to a first embodiment of the present invention.

Figure 2(a) is a timing diagram illustrating a test-mode operation of the input circuit of the first embodiment; and

Figure 2(b) is a timing diagram illustrating a normal-mode operation of the input circuit of the first embodiment.

Figure 3 is a block diagram illustrating a schematic arrangement of a data signal transmitting/receiving system including transmitter and receiver chips according to the first embodiment.

Figure 4 is a block diagram illustrating an internal configuration of the receiver chip.

Figure 5 is a block diagram illustrating a schematic arrangement of a data signal transmitting/receiving system in-

cluding transmitter and receiver chips according to a second embodiment of the present invention.

Figure 6 is a timing diagram how erroneous latching happens when the transition interval of a data signal from H into L level is long.

DESCRIPTION OF THE PREFERRED EMBODIMENTS

Hereinafter, preferred embodiments of the present invention will be described with reference to the accompanying drawings.

EMBODIMENT 1

Figures 1, 2, 3 and 4 illustrate a first exemplary embodiment of the present invention. Figure 3 illustrates an overall arrangement of a data signal transmitting/receiving system including transmitter and receiver chips.

As shown in Figure 3, the transmitter chip 10 includes a plurality of, or a number n of, output buffers 301, 302, ..., 30 n and an internal circuit 35. The internal circuit 35 sends out a mode-switching signal Con for switching a test mode into a normal operation mode, or vice versa, and data signals $D1$, $D2$, ..., Dn to the receiver chip 20 via these output buffers 301 through 30 n .

As shown in Figures 3 and 4, the receiver chip 20 includes: a number n of input circuits 401, 402, ..., 40 n ; an in-

ternal circuit **50**; a phase locked loop (PLL) circuit **51**; and a power supply circuit **52**. The input circuits **401** through **40n** respectively hold the data signals **D1**, **D2**, ..., **Dn** received. The data signals **D1''**, **D2''**, ..., **Dn''**, which are held in these input circuits **401** through **40n**, are then supplied to the internal circuit **50**. Responsive to a reference clock signal, the PLL circuit **51** generates a clock signal **CLK** and supplies the clock signal **CLK** to the internal circuit **50**. The power supply circuit **52** supplies a power supply voltage to the internal circuit **50** and a reference voltage **Vref** (described later) to the respective input circuits **401** through **40n**.

Figure 1 illustrates an internal configuration of the input circuit **401**. It should be noted that the other input circuits **402** through **40n** also have the same configuration as that of the input circuit **401** shown in Figure 1.

As shown in Figure 1, the input circuit **401** includes: an input buffer **11**; a delay section **60**; and a holding circuit **21**. The delay section **60** includes: a comparator **5**; two delay circuits **31** and **32**; and a selector **4**.

The input buffer **11** is a differential amplifier (differential input device), which receives the reference voltage **Vref** and the data signal **D1** and outputs an amplified signal **D1'** based on the difference between the voltage of the data signal **D1** and the reference voltage **Vref**. The reference voltage **Vref** is set equal to or lower than the power supply

voltage and higher than a ground potential. If the voltage of the input signal **D1** is higher than the reference voltage **Vref**, then the input buffer **11** outputs the power supply voltage as the "H-level" potential (i.e., logically high voltage).

5 Alternatively, if the voltage of the input signal **D1** is lower than the reference voltage **Vref**, then the input buffer **11** outputs the ground potential as the "L-level" potential (i.e., logically low voltage).

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10 The holding circuit **21** is implemented as a D-latch, which holds the amplified data signal **D1'** supplied from the input buffer **11** in synchronism with a leading edge of a clock signal **CLK2** delayed by the delay circuit **31** or **32**, and then outputs latched data signal **D1"**. That leading edge of the delayed clock signal **CLK2** is equivalent to the edge on which
15 the data signal is latched as defined in the appended claims.

The comparator **5** compares the phase of the amplified data signal **D1'** to that of the clock signal **CLK**. The result of phase comparison between a leading edge of the data signal **D1'** (i.e., transition from L into H level) and a corresponding leading edge of the clock signal **CLK** is output as a signal **Cde1F** or **Cde1B**. If the leading edge of the data signal **D1'** is prior to that of the clock signal **CLK**, then a pulse with a width equal to the phase difference therebetween is output as the signal **Cde1F**. Conversely, if the leading edge
20 of the data signal **D1'** is posterior to that of the clock sig-
25

the mode-switching signal **Con** is low.

If the data signal **D1'** is logically high, then the selector 4 selects the signal **CLK_LH** output from the delay circuit 31. Alternatively, if the data signal **D1'** is logically low, then the selector 4 selects the signal **CLK_HL** output from the delay circuit 32. Then, the selector 4 outputs the selected signal **CLK_LH** or **CLK_HL** as the clock signal **CLK2** to the holding circuit 21.

Next, the operation of the input circuit of the present invention will be described separately in terms of test and normal operation modes thereof.

Figures 2(a) and 2(b) are timing diagrams illustrating the operation of the input circuit in the test mode and the normal operation mode, respectively.

(Test mode)

First, a test mode interval (initializing interval) is provided to determine the respective amounts of time for which the clock signal **CLK** should be delayed.

When the signal **Con** reaches the H level, the initializing interval starts. At the outset of the initializing interval, the data signal **Data**, which repeatedly alternates its level between H and L, is output as test data from the transmitter chip 10 to the receiver chip 20. As shown in Figure 2(a), the length of the transition interval of the data sig-

nal **D1** from L into H is different from that of the transition interval of the data signal **D1** from H into L. Accordingly, the length of an interval during which the data signal **D1'** amplified by the input buffer **11** is high is not equal that of
5 an interval during which the signal **D1'** is low.

In this embodiment, in holding the data signal **D1'** responsive to a clock signal, a setup time **T1** of a sufficient length is determined in advance, which accounts for about 30% to about 50% of one cycle of the clock signal. Also, this
10 setup time **T1** is a relatively long period of time, which is longer than phase difference **T2** or **T3** described below. Suppose a leading edge of the data signal **D1'** turns out to be prior to an associated leading edge of the clock signal **CLK** as a result of phase comparison between the data signal **D1'**
15 and the clock signal **CLK** by the comparator **5**. Then, the delay circuit **31** determines the delay time **de1** as $T1 - T2$, where **T2** is a phase difference between these leading edges of the data signal **D1'** and the clock signal **CLK**. On the other hand, if a leading edge of the data signal **D1'** turns out to be pos-
20 terior to an associated leading edge of the clock signal **CLK**, then the delay circuit **31** determines the delay time **de1** as $T1 + T2$. Moreover, suppose a trailing edge of the data signal **D1'** turns out to be posterior to an associated leading edge of the clock signal **CLK**. Then, the delay circuit **32** deter-
25 mines the delay time **de2** as $T1 + T3$, where **T3** is a phase dif-

ference between the trailing edge of the data signal **D1'** and the leading edge of the clock signal **CLK**. Furthermore, if a trailing edge of the data signal **D1'** turns out to be prior to an associated leading edge of the clock signal **CLK**, then the
5 delay circuit **32** determines the delay time **de2** as $T1 - T3$. Accordingly, each leading edge of the delayed clock signal **CLK2** is always delayed from its associated leading or trailing edge of the data signal **D1'** for the setup time **T1**.

Once the delay time has been determined by the delay
10 circuit **31** or **32** in this manner, the delay circuit **31** or **32** delays the clock signal **CLK** for the determined amount of time, thereby outputting the delayed clock signal **CLK_LH** or **CLK_HL**. Specifically, one delayed clock signal **CLK_LH** is obtained by having the clock signal **CLK** delayed by the first delay circuit
15 **31**. The delay time thereof is determined based on the phase difference between a leading edge of the data signal **D1'** and an associated leading edge of the clock signal **CLK** during the transition of the data signal **D1'** from L into H level. On the other hand, the other delayed clock signal
20 **CLK_HL** is obtained by having the clock signal **CLK** delayed by the second delay circuit **32**. The delay time thereof is determined based on the phase difference between a trailing edge of the data signal **D1'** and an associated leading edge of the clock signal **CLK** during the transition of the data signal
25 **D1'** from H into L level.

The initializing interval terminates when this timing control is completed.

(Normal operation mode)

5 Next, when the mode-switching signal **Con** falls to the L level, the input circuit enters the normal operation mode.

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10 In this operation mode, a data signal is transferred as in an ordinary data transfer operation. The operation of holding a data signal in the holding circuit **21**, however, is performed responsive to a delayed clock signal **CLK2**, which is selected from the two delayed clock signals **CLK_HL** and **CLK_LH** derived during the initializing interval. The selector **4** selects one of the delayed clock signals **CLK_HL** and **CLK_LH** based on the logical state of the data signal **D1'**. Specifically,
15 cally, if the data signal **D1'** is currently logically high, then the selector **4** selects the signal **CLK_LH**. Alternatively, if the data signal **D1'** is logically low, then the selector **4** selects the signal **CLK_HL**. The signal selected by the selector **4** is supplied as the delayed clock signal **CLK2** to the
20 holding circuit **21**, which holds the data signal **D1'** in synchronism with the leading edge of the delayed clock signal **CLK2**.

25 By reducing a phase difference between a leading or trailing edge of the data signal **D1'** and an associated leading edge of the delayed clock signal **CLK2** in this manner, the

data signal **D1'** can be held correctly and erroneous latching can be prevented.

Next, the operation of the input circuit of the present invention as shown in Figure 2 and that of a conventional circuit, in which data is latched on every leading edge of an input clock signal **CLK** without using the selector 4 as shown in Figure 6, will be described in comparison. As shown in Figures 2 and 6, at a time **t0** synchronous with the first leading edge of the clock signal **CLK**, the data signal **D1'**, supplied from the differential amplifier 11, has a voltage higher than the reference voltage **Vref**, and is logically high. Thus, in Figure 6, if the data signal **D1'** is latched on the first leading edge of the clock signal **CLK**, then the data signal **D1"** is correctly latched in the holding circuit 21 as logically high. In Figure 2, the data signal **D1'** is latched on the first leading edge of the delayed clock signal **CLK_LH**, which is delayed from that of the clock signal **CLK** for a time **T1-T2**. However, the data signal **D1'** in the logically high state may be correctly latched on the first leading edge of the input clock signal **CLK** as in Figure 6. Thus, according to the present invention, two kinds of delay times, corresponding to the two logical states of the data signal **D1'** (i.e., H and L levels), do not always have to be calculated. That is to say, according to the present invention, only one of the delayed clock signals **CLK_LH** or **CLK_HL** may be selec-

tively produced for one of these two logical states. Responsive to this delayed clock signal, the data signal **D1'** in one of the two logical states may be latched. But the data signal **D1'** in the other logical state may be latched responsive to the original clock signal **CLK** without calculating the delay time for the other logical state. In such a case, the selector 4 receives the clock signal **CLK_LH** or **CLK_HL** delayed by one of the two delay circuits (e.g., 31) and the original clock signal **CLK**.

According to the present invention, the setup time **T1** does not always have to be provided. However, in order to latch the data signal **D1'** during a steady-state interval (in which the voltage is constant) after the transition interval of the data signal **D1'** is over, the setup time **T1** should preferably be provided.

Next, at a time **t1** synchronous with the next leading edge of the clock signal **CLK**, the data signal **D1'** supplied from the differential amplifier 11 is in the middle of level transition, but still has a voltage higher than the reference voltage **Vref**.

That is to say, the data signal **D1'** is logically high at the time **t1** in both Figures 2 and 6. Thus, in Figure 6, if the data signal **D1'** is latched at the time **t1** synchronous with the second leading edge of the clock signal **CLK**, the holding circuit 21 holds the H-level data signal **D1'**, not the correct L-level data signal **D1'** as indicated by the broken line. As a

result, erroneous data is latched. In contrast, in this embodiment, the L-level data signal **D1'** is latched correctly by the holding circuit **21** at a time **t2** synchronous with the leading edge of the delayed clock signal **CLK_HL** obtained by delaying the clock signal **CLK** for the predetermined amount of time **de2** ($= T1+T3$) as shown in Figure 2. That is to say, the data signal **D1'** is latched after the voltage of the data signal **D1'** has fallen to be lower than the reference voltage **Vref**. As a result, the present invention can prevent erroneous latching.

In this embodiment, the data signal **D1'** is supposed to be in the same logical state during one cycle of the clock signal, i.e., the frequency of the data signal is supposed to be one half as high as that of the clock signal. However, the application of the present invention is not limited to such a situation. Thus, the applicable frequencies of the data and clock signals are not particularly limited according to the present invention.

In the foregoing embodiment, the data signal **D1'** is supposed to be latched on the leading edges of the clock signal. Alternatively, the data signal may also be latched either on the trailing edges of the clock signal or on the leading and trailing edges thereof in combination.

Furthermore, in the foregoing embodiment, the first or second delay circuit **31** or **32** is selectively used. Specifically, the clock signal delayed by the first delay circuit **31**

is supposed to be selected when the data signal is logically high, while the clock signal delayed by the second delay circuit 32 is supposed to be selected when the data signal is logically low. However, if it is clear that only the leading edges of the data signal are posterior to associated leading edges of the clock signal, then the phase may be controlled at only these leading edges of the data signal D1' by using only the first delay circuit 31. Also, if it is clear that only the trailing edges of the data signal are posterior to associated leading edges of the clock signal, then the phase may be controlled at only these trailing edges of the data signal D1' by using only the second delay circuit 32. In any of these cases, the selector 4 is not necessary. Moreover, in the foregoing description, the setup time T1 is supposed to be constant if the time is long enough to hold the data. However, if the time T1 is varied responsive to another external control signal, the input circuit can be modified to satisfy the demand of even higher-speed operation.

EMBODIMENT 2

Next, a second exemplary embodiment of the present invention will be described with reference to Figure 5.

In Figure 5, a transmitter chip 70 (equivalent to the output circuit of the present invention as defined in the appended claims) transmits a plurality of data signals to be

received by a receiver chip **80**. It should be noted that only one data signal **D1** is illustrated in Figure 5 for the sake of simplicity. The receiver chip **80** includes the internal circuit **50**, PLL circuit **51** and power supply circuit **52**, all of which have the same configurations as the counterparts illustrated in Figure 3. The receiver chip **80** further includes a plurality of input circuits (though only one input circuit **81** is illustrated in Figure 5 for the sake of simplicity). In this embodiment, although not shown in any of the drawings, each of these input circuits includes only the differential amplifier **11**, comparator **5** and holding circuit **21** of the input circuit **401** shown in Figure 1. That is to say, the input circuit does not include the first and second delay circuits **31** and **32** and the selector **4**. Responsive to the clock signal **CLK** supplied from the PLL circuit **51**, the holding circuit **21** latches the data signal **D1'** supplied from the differential amplifier **11**.

On the other hand, the transmitter chip **70** includes a plurality of output buffers each transmitting an associated data signal (although only one output buffer **71** for outputting the data signal **D1** is illustrated in Figure 5 for the sake of simplicity). Each of these output buffers has the same internal configuration. For example, the output buffer **71** includes three p-channel transistors **TP1**, **TP2** and **TP3**, connected to the power supply, for turning the logical volt-

age level of the data signal **D1** into high, and three n-channel transistors **TN1**, **TN2** and **TN3** grounded for turning the logical voltage level of the data signal **D1** into low. The transmitter chip **70** further includes first and second controllers **72** and **73** for controlling the p-channel transistors **TP1**, **TP2** and **TP3** and n-channel transistors **TN1**, **TN2** and **TN3**, respectively. The first controller **72** receives the signal **Cde1F** or **Cde1B** representing the result of comparison between a leading edge of the data signal **D1'** and an associated leading edge of the clock signal **CLK**, i.e., the signals representing the length of the transition interval of the data signal, from the comparator **5** in the input circuit **81**. On the other hand, the second controller **73** receives the signal **Cde2F** or **Cde2B** representing the result of comparison between a trailing edge of the data signal **D1'** and an associated leading edge of the clock signal **CLK**, from the comparator **5** in the input circuit **81**. These signals **Cde1F/Cde1B** and **Cde2F/Cde2B** are shown in Figure 2. Suppose the signal **Cde1B** is supplied to the first controller **72**, i.e., the time when the voltage of the data signal **D1'** rises to reach the reference voltage **Vref** is later than the leading edge of the clock signal **CLK** during the transition of the data signal **D1'** from L to H level. In other words, if the transition interval is relatively long, then the first controller **72** increases the number of p-channel transistors to be turned ON, thereby en-

hancing the total drivability of the output buffer 71 to shorten the transition interval. On the other hand, suppose the signal **Cde2B** is supplied to the second controller 73, i.e., the time when the voltage of the data signal **D1'** drops to reach the reference voltage **Vref** is later than the leading edge of the clock signal **CLK** during the transition of the data signal **D1'** from H to L level. In other words, if the transition interval is relatively long, then the second controller 73 also increases the number of n-channel transistors to be turned ON, thereby enhancing the total drivability of the output buffer to shorten the transition interval.

The data signal **D1'** is fed back to the first and second controllers 72 and 73. Accordingly, even if these controllers 72 and 73 do not receive the signals **Cde1F/Cde1B** and **Cde2F/Cde2B** from the comparator 5, the drivability of the output buffer 71 can be known based on the feedback signal. And if the drivability is low, the number of transistors to be turned ON may be increased.

Accordingly, in this embodiment, the drivability of the output buffer 71 can be controlled on the transmitting end, i.e., in the transmitter chip 70, thereby adjusting the length of the transition interval of the data signal **D1'** from H into L level or vice versa. Thus, as in the first embodiment, skewing can be suppressed and a high-speed operation is ensured in a frequency region where the clock frequency is

several hundreds megahertz or more.

In the foregoing discussion, the present invention has been described as being applied to an input circuit receiving a single data signal **D1'**. However, the present invention is
5 also applicable to an input circuit receiving differential input signals. In such a case, these differential signals are supplied to the differential amplifier **11**.

WHAT IS CLAIMED IS:

1. An input circuit comprising:

delay means for defining a delay time for at least one logical state of a data signal and thereby delaying the clock signal for the delay time defined; and

a holding circuit for holding the data signal responsive to the delayed clock signal.

2. The input circuit of Claim 1, wherein the delay means defines the delay time such that an edge of the clock signal, on which the data signal is intended to be latched and which is included within a transition interval of the data signal, is delayed to a point in time after the transition interval of the data signal is over.

3. The input circuit of Claim 2, wherein the delay means comprises:

a comparator for comparing the edge of the clock signal, on which the data signal is intended to be latched, to at least one of leading and trailing edges of the data signal; and

a delay circuit for defining the delay time based on a result of comparison performed by the comparator.

4. The input circuit of Claim 2, wherein the delay means comprises:

a comparator for comparing the edge of the clock signal, on which the data signal is intended to be latched, to lead-

ing and trailing edges of the data signal;

a first delay circuit for defining the delay time for a logically high state of the data signal based on a result of comparison, performed by the comparator, between one of the leading edges of the data signal and the edge of the clock signal;

a second delay circuit for defining the delay time for a logically low state of the data signal based on a result of comparison, performed by the comparator, between one of the trailing edges of the data signal and the edge of the clock signal; and

a selector for selecting the delay time defined by the first delay circuit when the data signal is in the logically high state or the delay time defined by the second delay circuit when the data signal is in the logically low state.

5. The input circuit of Claim 3 or 4, wherein the delay circuit defines the delay time based on the result of comparison performed by the comparator and a setup time for correctly latching the data signal.

6. An output circuit comprising:

a driver including a plurality of devices outputting a data signal, the total drivability of the devices being controllable; and

a controller, responsive to a signal representing a transition interval length of the data signal, for increasing

or decreasing the drivability of the driver.

7. The output circuit of Claim 6, wherein the controller receives the signal, representing the transition interval length of the data signal, from an input circuit to which the data signal is output from the output circuit.

ABSTRACT OF THE DISCLOSURE

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An input circuit includes: a comparator; first and second delay circuits; a selector; an input buffer; and a holding circuit. The comparator compares the leading and/or trailing edges of a data signal, supplied from the input buffer, to an edge of a clock signal on which the data signal is intended to be latched. Based on the results of the comparison, the first and second delay circuits delay the clock signal for respectively predetermined amounts of time. If the data signal is logically high, then the selector selects a delayed clock signal supplied from the first delay circuit. Alternatively, if the data signal is logically low, then the selector selects another delayed clock signal supplied from the second delay circuit. Then, the delayed clock signal, selected by the selector, is latched in the holding circuit. The input circuit with such a configuration prevents skewing from being caused by a difference in length between the transition interval of the data signal from H into L level and that of the data signal from L into H level. As a result, data can be transferred at a much higher speed even if the clock frequency is very high.

Fig. 1

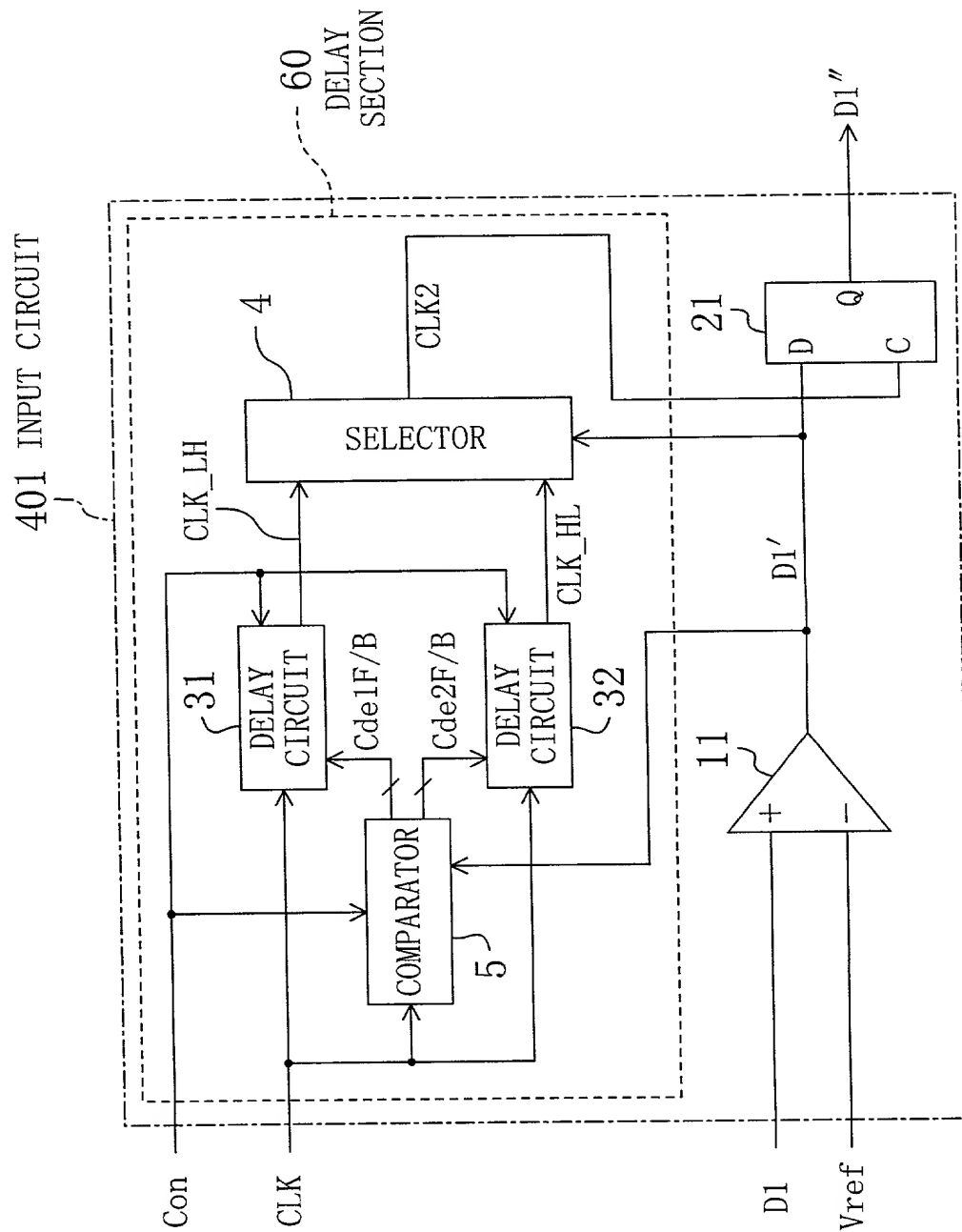


Fig. 2(a) Con=H

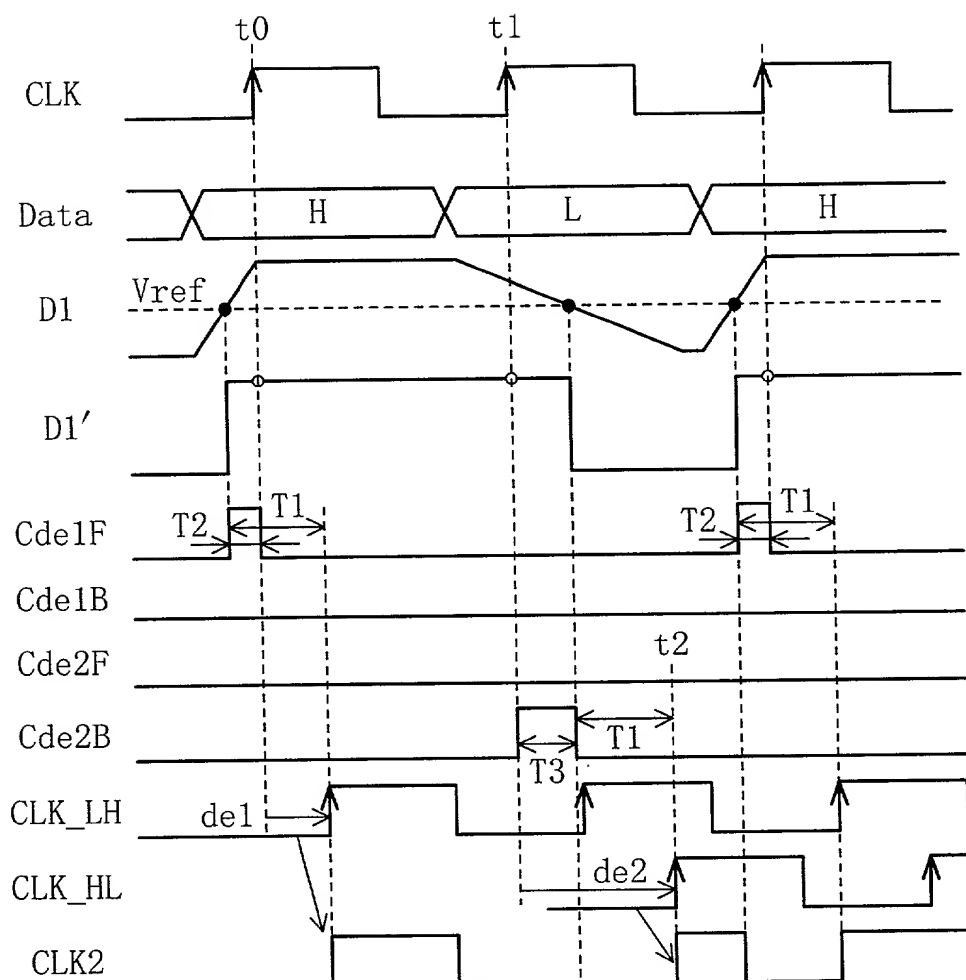


Fig. 2(b) Con=L

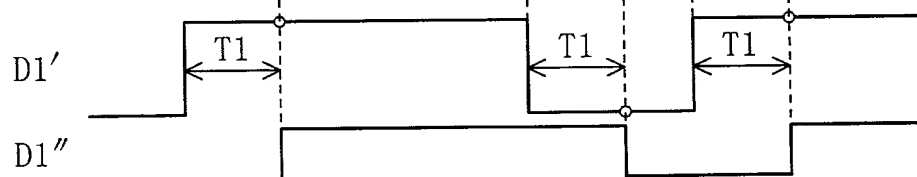


Fig. 3

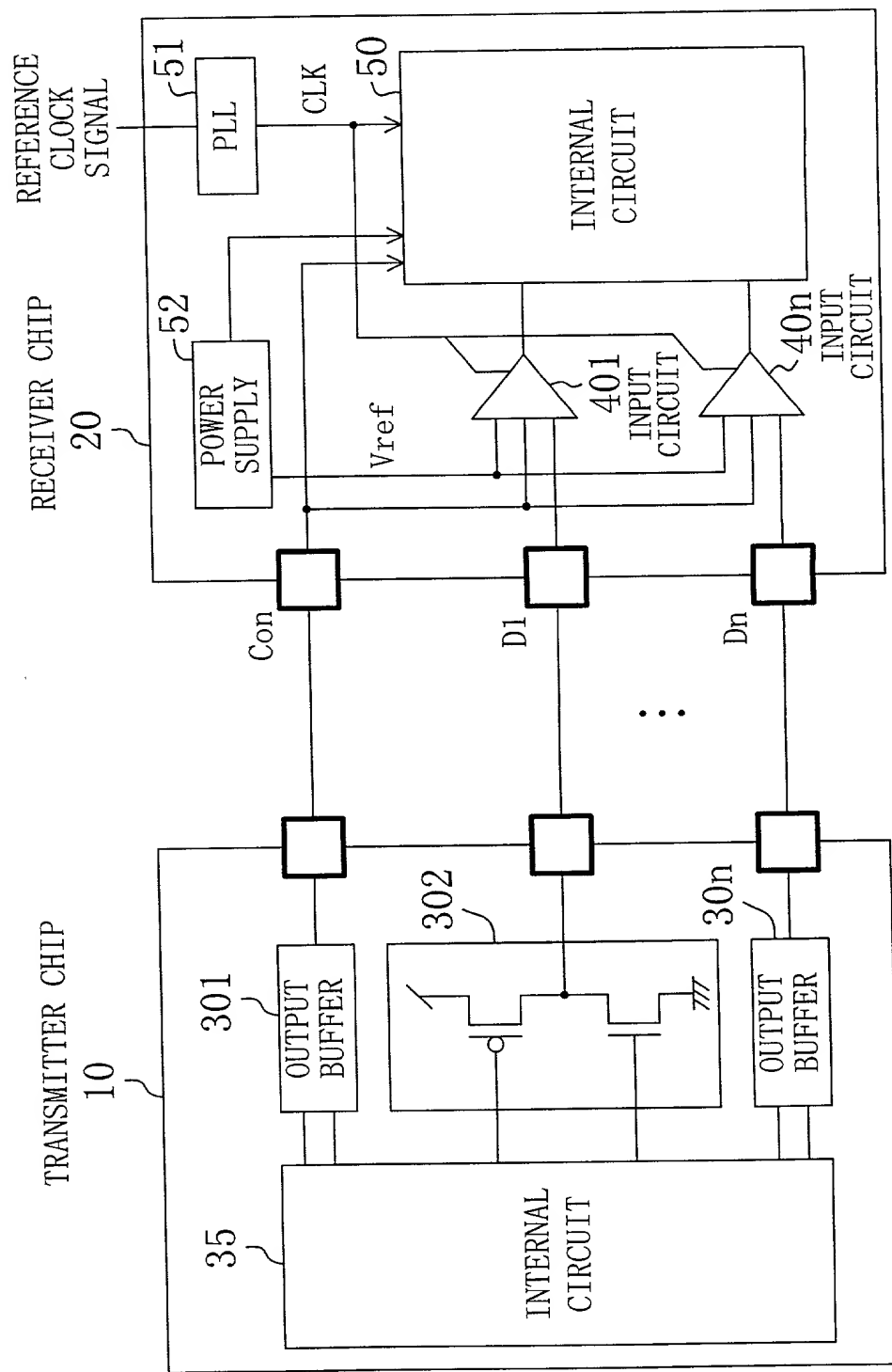


Fig. 4

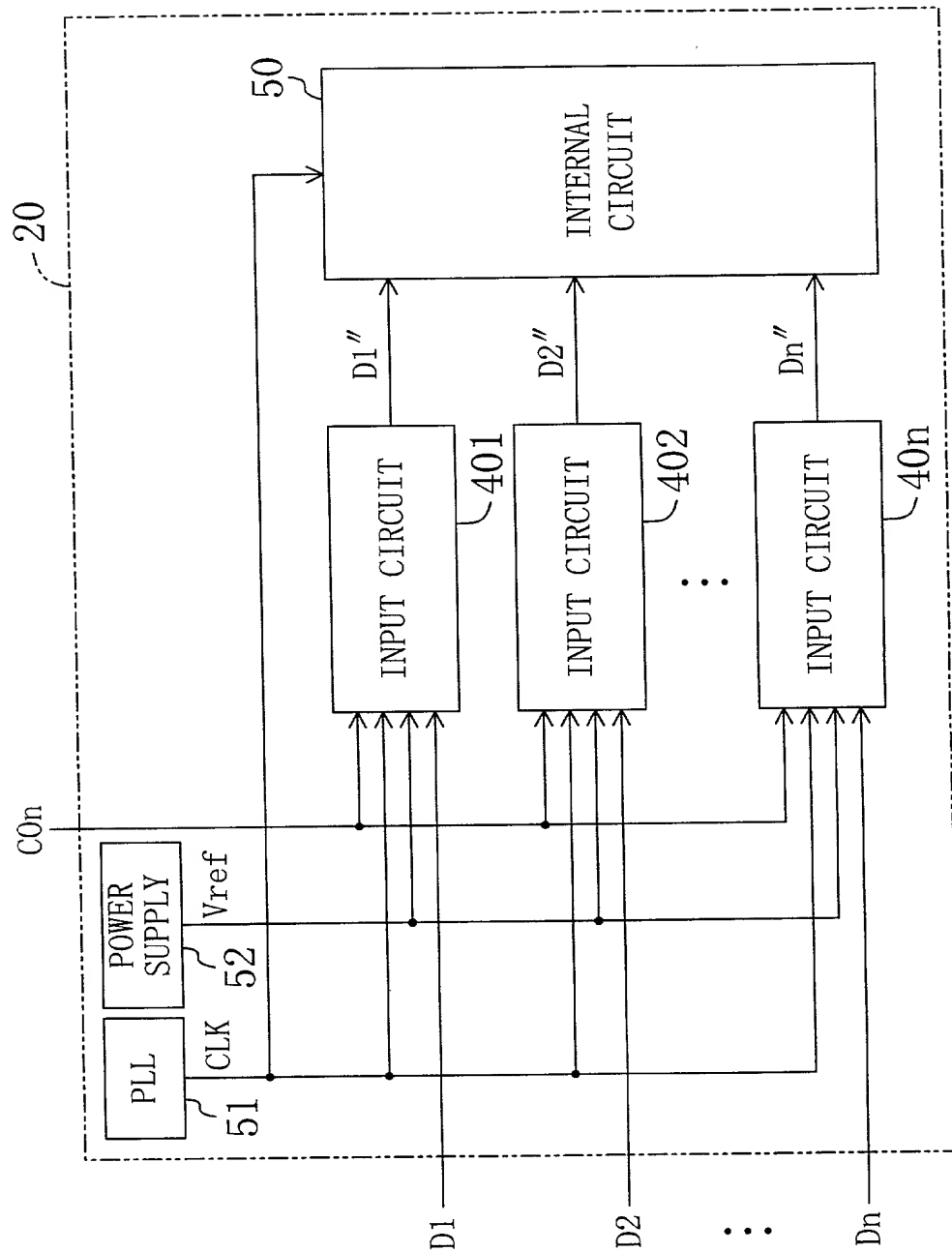


Fig. 5

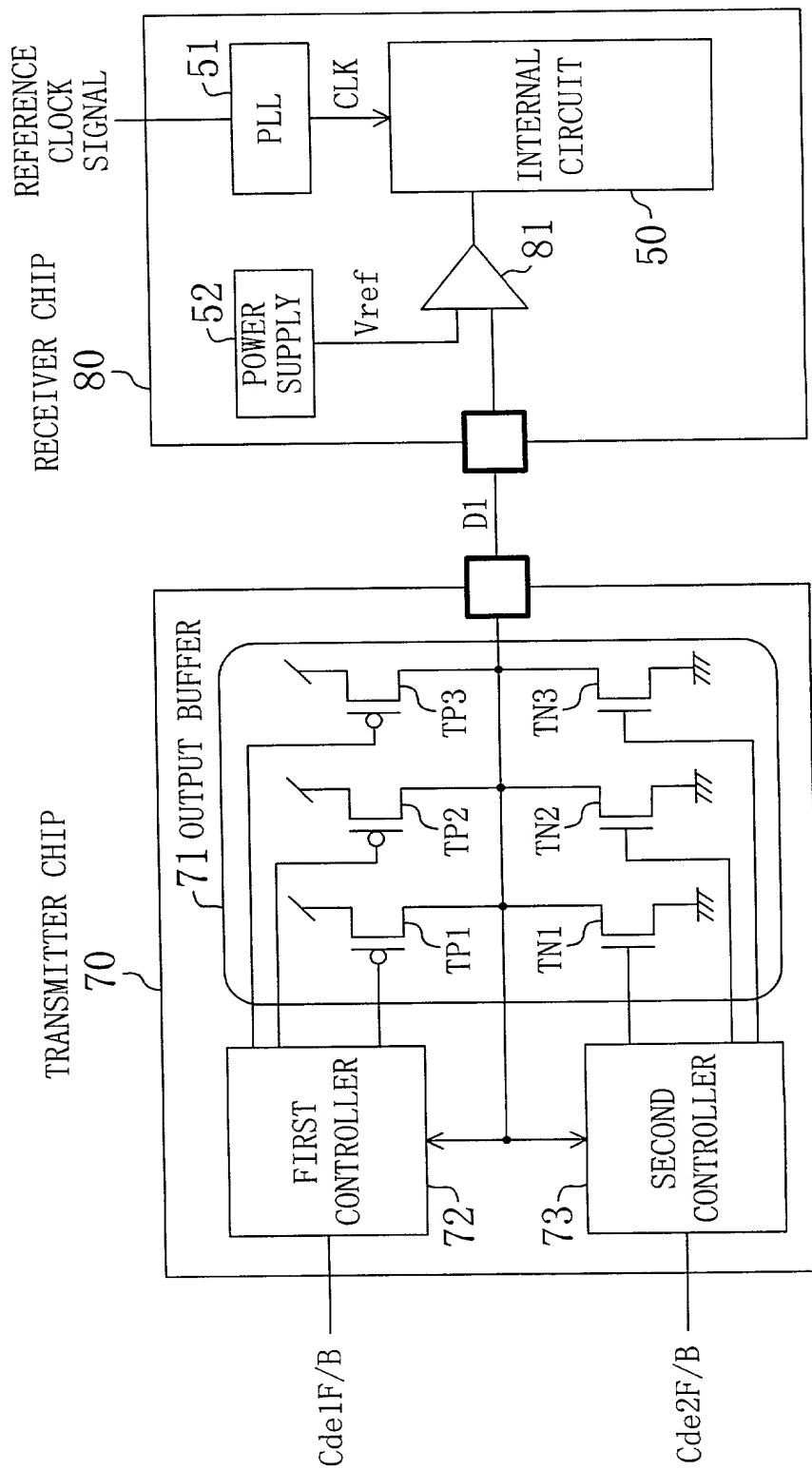
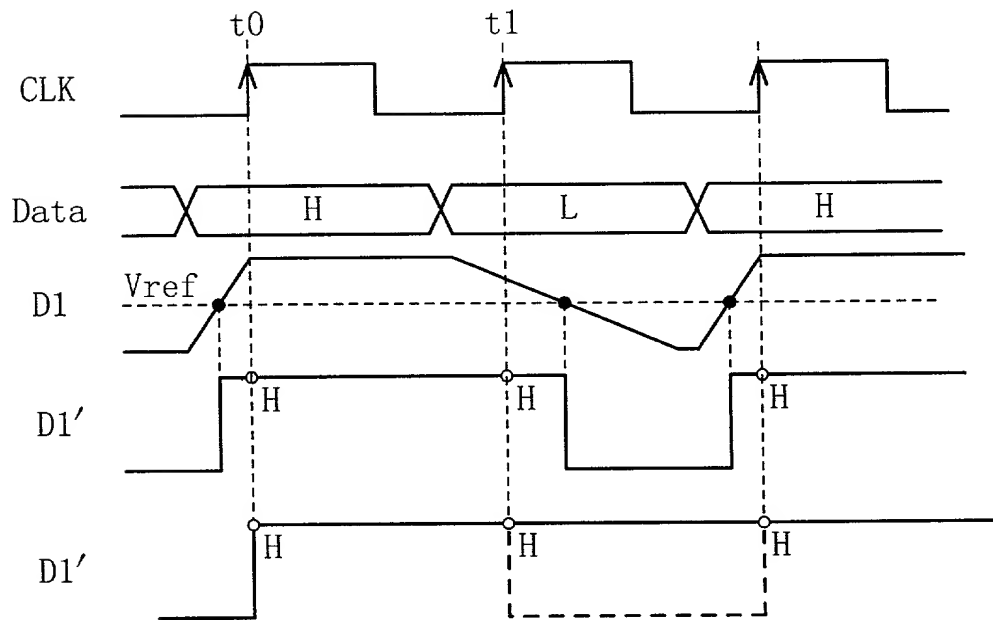


Fig. 6
Prior Art



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COMBINED DECLARATION/POWER OF ATTORNEY
FOR PATENT APPLICATION

As a below named inventor, I hereby declare that:

My residence, post office address and citizenship are as stated below next to my name.

I believe I am the original, first and sole inventor (if only one name is listed below) or an original, first and joint inventor (if plural names are listed below) of the subject matter which is claimed and for which a patent is sought on the invention entitled
INPUT CIRCUIT AND OUTPUT CIRCUIT

_____, the specification of which

(check one) X is attached hereto.

_____ was filed on _____ as
 Application Serial No. _____.

I hereby state that I have reviewed and understand the contents of the above identified specification, including the claims, as amended by any amendment referred to above.

I acknowledge the duty to disclose information which is material to the examination of this application in accordance with Title 37, Code of Federal Regulations, § 1.56(a).

I hereby claim foreign priority benefits under Title 35, United States Code, § 119 of any foreign application(s) for patent or inventor's certificate listed below and have also identified below any foreign application for patent or inventor's certificate having a filing date before that of the application on which priority is claimed:

Prior Foreign Application(s)	Priority Claimed
<u>10-118349</u> (Number)	<u>JAPAN</u> (Country)
<u>28/04/1998</u> (Day/Month/Year Filed)	<u>X</u> Yes ___ No
_____ (Number)	_____ (Country)
_____ (Day/Month/Year Filed)	___ Yes ___ No
_____ (Number)	_____ (Country)
_____ (Day/Month/Year Filed)	___ Yes ___ No

I hereby claim the benefit under Title 35, United States Code, § 120 of any United States application(s) listed below and, insofar as the subject matter of each of the claims of this application is not disclosed in the prior United States application in the manner provided by the first paragraph of Title 35, United States Code, § 112, I acknowledge the duty to disclose material information as defined in Title 37, Code of Federal Regulations, § 1.56(a) which occurred between the filing date of the prior application and the national or PCT international filing date of this application:

65240 659660

(Appln. Serial No.)	(Filing Date)	(Status-patented, pending, abandoned)
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(Appln. Serial No.)	(Filing Date)	(Status-patented, pending, abandoned)
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I hereby appoint as my attorneys, with full power of substitution and revocation, to prosecute the patent application identified above and to transact all business in the U.S. Patent and Trademark Office connected therewith: Raphael V. Lupo (Reg. No. 28,363); Jack Q. Lever, Jr. (Reg. No. 28,149); Kenneth L. Cage (Reg. No. 26,151); Stanislaus Aksman (Reg. No. 28,562); Paul Devinsky (Reg. No. 28,553); Edward E. Kubasiewicz (Reg. No. 30,020), Michael E. Fogarty (Reg. No. 36,139); Brian E. Ferguson (Reg. No. 36,801); Robert W. Zelnick (Reg. No. 36,976); and Wilhelm F. Gadiano (Reg. No. 37,136).

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PATENT AND TRADEMARK OFFICE

The undersigned hereby authorizes the U.S. attorneys named herein to accept and follow instructions from Maeda Patent Office as to any action to be taken in the Patent and Trademark Office regarding this application without direct communication between the U.S. attorney and the undersigned. In the event of a change in the persons from whom instructions may be taken, the U.S. attorneys named herein will be so notified by the undersigned.

I hereby declare that all statements made herein of my own knowledge are true and that all statements made on information and belief are believed to be true; and further that these statements were made with the knowledge that willful false statements and the like so made are punishable by fine or imprisonment, or both, under Section 1001 of Title 18 of the United States Code and that such willful false statements may jeopardize the validity of the application or any patent issued thereon.

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